Chapter 4 THE Z TRANSFORM 4.9 Spectral Representation of Discrete-Time Signals

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Spectral Representation of Discrete-Time Signals

▶ The *frequency spectrum* of a discrete-time signal is given by

$$X(z)\big|_{z=e^{j\omega T}}=X(e^{j\omega T})$$

and it is a complex quantity in general.

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• The magnitude and angle of $X(e^{j\omega T})$, i.e.,

$$A(\omega) = |X(e^{j\omega T})|$$
 and $\phi(\omega) = \arg X(e^{j\omega T})$

define the *amplitude spectrum* and *phase spectrum* of the discrete-time signal x(nT), respectively.

Spectral Representation Cont'd

The exponential function e^{jωT} is a complex number of magnitude 1 and angle ωT and as ω is increased from zero to 2π/T, e^{jωT} will trace a circle of radius 1 in the z plane, which is referred to as the *unit circle*.

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Spectral Representation Cont'd

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- ► In effect, the frequency spectrum of a discrete-time signal, x(nT), can be deduced by evaluating its z transform, X(z), on the unit circle.

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Geometrical Features of Z Plane

▶ If $\omega = 0$, then $e^{j\omega T} = e^0 = 1$, i.e., point A corresponds to zero frequency.



► At half the sampling frequency, $\omega = \omega_s/2 = \pi/T$ and hence $e^{j\omega T} = e^{j\pi} = -1$, i.e., point C corresponds to the Nyquist frequency.



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At the sampling frequency, ω = ω_s = 2π/T and hence e^{jωT} = e^{j2π} = 1, i.e., point A also corresponds to the sampling frequency.



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► If vector e^{jωT} is rotated k complete revolution starting from some arbitrary point B, it will return to point B.



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Hence

$$e^{j(\omega T+2\pi k)} = e^{(j\omega+2\pi k/T)T} = e^{(j\omega+k\omega_s)T} = e^{j\omega T}$$

and, therefore,

$$X(e^{j(\omega+k\omega_s)T}) = X(e^{j\omega T})$$

In effect, the frequency spectrum of a discrete-time signal is a periodic function of frequency with period ω_s .

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Example

Obtain expressions for the frequency, amplitude, and phase spectrums of the signal

$$x(nT) = u(nT)e^{-lpha nT}\sin\omega_0 nT$$

Solution The *z* transform of the signal can be obtained from Table 3.2 as

$$X(z) = \frac{ze^{-\alpha T} \sin \omega_0 T}{z^2 - 2ze^{-\alpha T} \cos \omega_0 T + e^{-2\alpha T}}$$

X(z) can be expressed as

$$X(z) = \frac{a_1 z}{z^2 + b_1 z + b_0}$$

where

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$$a_1 = e^{-\alpha T} \sin \omega_0 T \quad b_0 = e^{-2\alpha T} \quad b_1 = -2e^{-\alpha T} \cos \omega_0 T$$

 $X(z) = \frac{a_1 z}{z^2 + b_1 z + b_0}$

Hence the frequency spectrum of the signal is given by

$$X(e^{j\omega T}) = \frac{a_1 e^{j\omega T}}{e^{j2\omega T} + b_1 e^{j\omega T} + b_0}$$

=
$$\frac{a_1 e^{j\omega T}}{\cos 2\omega T + j \sin 2\omega T + b_1 \cos \omega T + jb_1 \sin \omega T + b_0}$$

=
$$\frac{a_1 e^{j\omega T}}{b_0 + b_1 \cos \omega T + \cos 2\omega T + j(b_1 \sin \omega T + \sin 2\omega T)}$$

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Example Cont'd

The amplitude and phase spectrums can be deduced by letting

$$X(e^{j\omega T}) = \frac{a_1 e^{j\omega T}}{b_0 + b_1 \cos \omega T + \cos 2\omega T + j(b_1 \sin \omega T + \sin 2\omega T)}$$
$$= A(\omega)e^{j\phi(\omega)}$$

Hence

$$A(\omega) = \frac{|a_1| \cdot |e^{j\omega T}|}{|(b_0 + b_1 \cos \omega T + \cos 2\omega T) + j(b_1 \sin \omega T + \sin 2\omega T)|}$$

= $\frac{|a_1|}{\sqrt{(b_0 + b_1 \cos \omega T + \cos 2\omega T)^2 + (b_1 \sin \omega T + \sin 2\omega T)^2}}$
= $\frac{|a_1|}{\sqrt{1 + b_0^2 + b_1^2 + 2b_1(1 + b_0)\cos \omega T + 2b_0\cos 2\omega T}}$

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Example Cont'd

$$\phi(\omega) = \arg(a_1) + \arg e^{j\omega T} - \arg[b_0 + b_1 \cos \omega T + \cos 2\omega T + j(b_1 \sin \omega T + \sin 2\omega T)]$$

= $\arg a_1 + \omega T - \tan^{-1} \frac{b_1 \sin \omega T + \sin 2\omega T}{b_0 + b_1 \cos \omega T + \cos 2\omega T}$

where

$$\arg a_1 = egin{cases} 0 & ext{if } a_1 \geq 0 \ -\pi & ext{otherwise} \end{cases}$$

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Amplitude spectrum

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This slide concludes the presentation. Thank you for your attention.